

## 1A High Side Driver with Overload Protection

December 1997

### Features

- Over Operating Temperature Range . . -40°C to 125°C
  - Max  $V_{SAT}$  at 1A . . . . . 1V
  - Current Switching Capability . . . . . 1A
  - Power Supply Range . . . . . 4.5V to 25V
- Over-Voltage Shutdown Protected
- Over-Current Limiting
- Thermal Limiting Protection
- Negative Output Voltage Clamp
- CMOS/TTL Logic Level Control Input
- Load Dump . . . . . 60V<sub>PEAK</sub>
- Reverse Battery Protection to -16V

### Applications

- Motor Driver/Controller
- Driver for Solenoids, Relays and Lamps
- MOSFET and IGBT Driver
- Driver for Temperature Controller

### Ordering Information

PART NUMBER	TEMP. RANGE (°C)	PACKAGE	PKG. NO.
HIP1030AS	-40 to 125	5 Ld TS-001AA SIP	Z5.067C

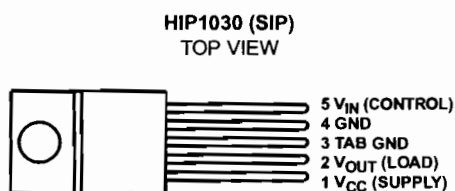
### Description

The HIP1030 is a High Side Driver Power Integrated Circuit designed to switch power supply voltage to an output load. It is the equivalent of a PNP pass transistor operated as a protected high side current switch in the saturated ON state with low forward voltage drop at the maximum rated current. The HIP1030 has low output leakage and low idle current in the OFF state.

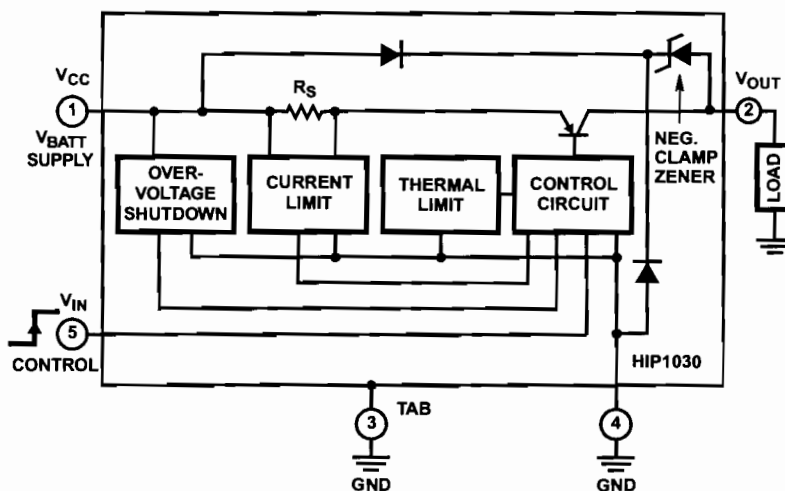
The Functional Block Diagram for the HIP1030 shows the protection control circuit functions of over-current, overvoltage and over-temperature. A small metal resistor senses over-current in the power supply path of the pass transistor and load. Overvoltage detection and shutdown of the output driver occurs when a comparator determines that the supply voltage has exceeded a comparator reference level. Over-temperature is sensed from a  $V_{BE}$  differential sense element that is thermally close to the output drive transistor. In addition to the input detected overvoltage protection, negative peak voltage of a switched inductive load is clamped with an internal zener diode. An internal bandgap voltage source provides a stable voltage reference over the operating temperature range, providing bias and reference control for the protection circuits.

The HIP1030 is particularly well suited for driving lamps, relays, and solenoids in automotive and industrial control applications where voltage and current overload protection at high temperatures is required. The HIP1030 is supplied in a 5 lead TS-001AA Power SIP package.

### Pinout



### Functional Block Diagram



# HIP1030

## Absolute Maximum Ratings

Maximum Supply Voltage  $V_{CC}$  . . . See O.V. Shutdown Limit,  $V_{OVSD}$   
 Input Voltage,  $V_{IN}$  (Note 1) . . . . . -1V to ( $V_{CC} - 0.5V$ )  
 Load Current,  $I_{OUT}$  . . . . . Internal Limiting  
 Load Dump (Survival) . . . . .  $\pm 60V_{PEAK}$   
 Reverse Battery . . . . . -16V

## Operating Conditions

Temperature Range . . . . . -40°C to 125°C

## Thermal Information

Thermal Resistance (Typical, Note 2)  $\theta_{JA}$  (°C/W)  $\theta_{JC}$  (°C/W)  
 Plastic SIP Package . . . . . 50 4  
 Maximum Power Dissipation (Note 3)  
 At  $T_A = 125^\circ\text{C}$ , Infinite Heat Sink . . . . . 6.25W  
 Maximum Junction Temperature,  $T_J$  . . . . . 150°C  
 Maximum Storage Temperature Range . . . . . -40°C to 150°C  
 Maximum Lead Temperature (Soldering 10s) . . . . . 300°C  
 (Lead Tips Only)

**CAUTION:** Stresses above those listed in "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress only rating and operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied.

## NOTES:

1. The Input Control Voltage,  $V_{IN}$  shall not be greater than ( $V_{CC} - 0.5V$ ) and shall not exceed +7V when  $V_{CC}$  is greater than 7.5V.
2.  $\theta_{JA}$  is measured with the component mounted on an evaluation PC board in free air.
3. The worst case thermal resistance,  $\theta_{JC}$  for the SIP TS-001AA 5 lead package is 4°C/W. The calculation for dissipation and junction temperature rise due to dissipation is:

$$P_D = (V_{CC} - V_{OUT})(I_{OUT}) + (V_{CC})(I_{CCMAX} - I_{OUT}) \text{ or } (V_{CC})(I_{CCMAX}) - (V_{OUT})(I_{OUT})$$

$$T_J = T_{AMBIENT} + (P_D)(\theta_{JC}) \text{ for an infinite Heat Sink.}$$

Refer to Figure 1 for Derating based on Dissipation and Thermal Resistance. Derating from 150°C is based on the reciprocal of thermal resistance,  $\theta_{JC} + \theta_{HS}$ . For example: Where  $\theta_{JC} = 4^\circ\text{C/W}$  and given  $\theta_{HS} = 6^\circ\text{C/W}$  as the thermal resistance of an external Heat Sink, the junction-to-air thermal resistance,  $\theta_{JA} = 10^\circ\text{C/W}$ . Therefore, for the maximum allowed dissipation, derate 0.1W/°C for each degree from  $T_{AMB}$  to the maximum rated junction temperature of 150°C. If  $T_{AMB} = 100^\circ\text{C}$ , the maximum  $P_D$  is  $(150 - 100) \times 0.1\text{W/}^\circ\text{C} = 5\text{W}$ .

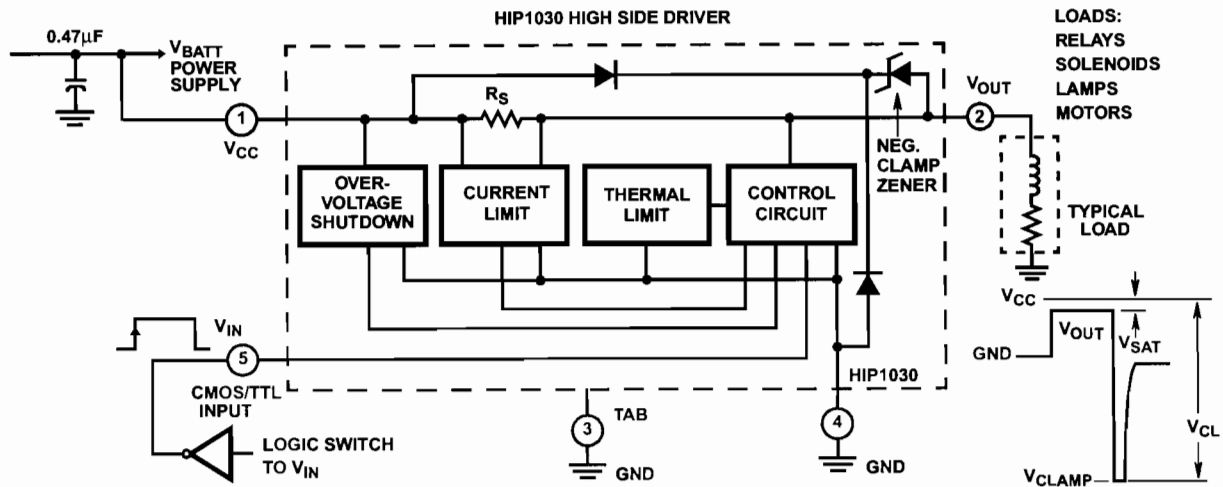
## Electrical Specifications $T_A = -40^\circ\text{C}$ to $125^\circ\text{C}$ , $V_{IN} = 2V$ , $V_{CC} = +12V$ , Unless Otherwise Specified

PARAMETER	SYMBOL	TEST CONDITIONS	MIN	TYP	MAX	UNITS
Operating Voltage Range	$V_{CC}$		4.5	-	25	V
Over-Voltage Shutdown	$V_{OVSD}$	$R_L = 1K\Omega$ ; $V_{IN} = 2V$	26	33	38	V
Over-Temperature Limiting	$T_{SD}$		-	150	-	°C
Negative Pulse Output Clamp Voltage	$V_{CL}$	$I_{CL} = -100\text{mA}$ ; $V_{CC} = 4.5V$ to $25V$	( $V_{CC} - 35$ )	( $V_{CC} - 30.5$ )	( $V_{CC} - 28$ )	V
Short Circuit Current Limiting	$I_{SC}$	(Note 4)	1.1	1.6	2.5	A
Input Control ON	$V_{IH}$		2.0	-	-	V
Input Control OFF	$V_{IL}$		-	-	0.8	V
Input Current High	$I_{IH}$	$V_{IN} = 5.5V$ , $V_{CC} = 6V$ to $24V$	6	-	40	$\mu\text{A}$
Input Current Low	$I_{IL}$	$V_{IN} = 0.8V$ , $V_{CC} = 6V$ to $24V$	6	-	30	$\mu\text{A}$
Supply Current, Full Load Input Control ON	$I_{CCMAX}$	$V_{IN} = 2V$ ; $I_{OUT} = 1.0A$ ;	-	1.05	1.1	A
Supply Current, No Load Input Control OFF	$I_{CCMIN}$	$V_{IN} = 0V$ ; $I_{OUT} = 0A$ ;	-	55	100	$\mu\text{A}$
Input-Output Forward Voltage Drop ( $V_{CC} - V_{OUT}$ )	$V_{SAT}$	$I_{OUT} = 1A$ ; $V_{CC} = 4.5V$ to $25V$	-	0.6	1	V
Output Leakage	$I_{OUT\_LK}$	$V_{IN} = 0.8V$ ; $V_{CC} = 6V$ to $24V$	-	4	50	$\mu\text{A}$
Turn ON Time	$t_{ON}$	$R_L = 80\Omega$ ; (Note 5)	-	5	20	$\mu\text{s}$
Turn OFF Time	$t_{OFF}$	$R_L = 80\Omega$ ; (Note 5)	-	25	65	$\mu\text{s}$

## NOTES:

4. Short circuit current will be reduced when thermal shutdown occurs. Testing of short circuit current may require a short duration pulse. See Figure 7.
5. Refer to Figures 3A and 3B for typical switching speeds with a 20 $\Omega$  Load.

## Typical Applications



## Typical Performance Curves

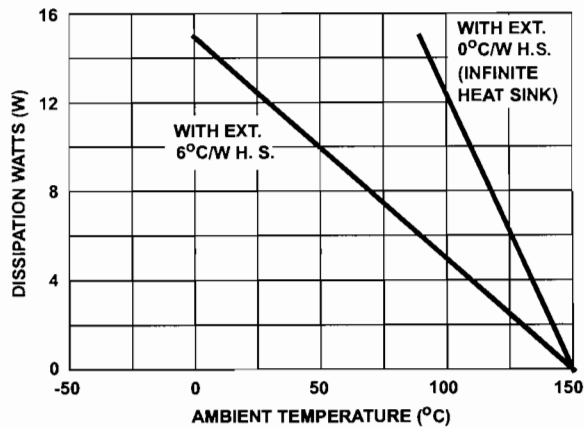


FIGURE 1. DISSIPATION DERATING CURVES

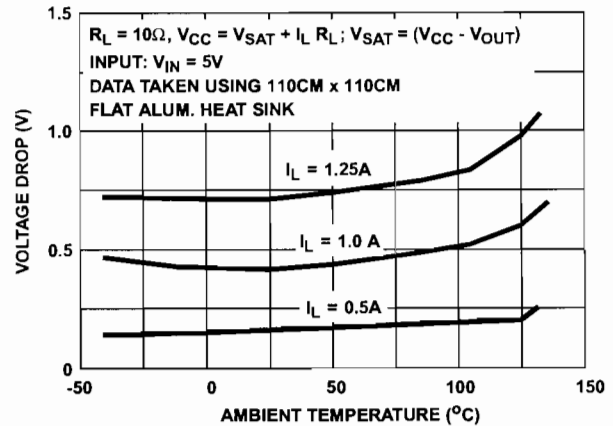


FIGURE 2. TYPICAL FORWARD VOLTAGE DROP,  $V_{SAT}$  CHARACTERISTICS vs AMBIENT OPERATING TEMPERATURE

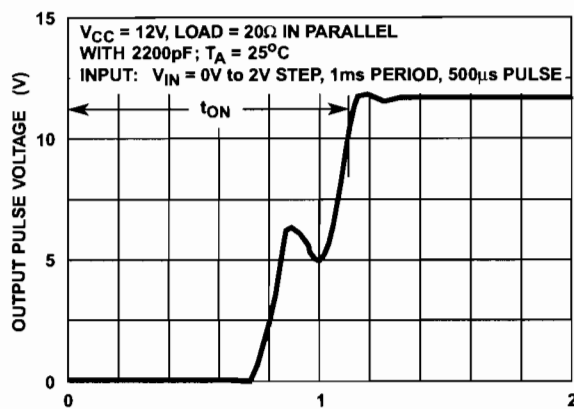


FIGURE 3A. OUTPUT TURN-ON TIME ( $\mu s$ )

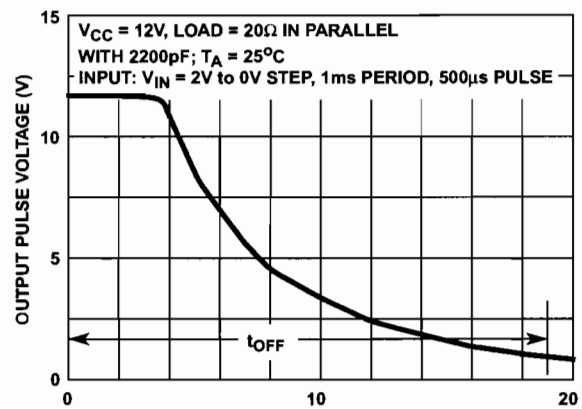


FIGURE 3B. OUTPUT TURN-OFF TIME ( $\mu s$ )

FIGURE 3. TYPICAL RISE TIME AND FALL TIME CHARACTERISTICS OF THE HIP1030 WITH A RESISTIVE AND CAPACITIVE LOAD. THE TURN-ON TIME OF APPROXIMATELY  $1.1\mu s$  IS PRIMARILY DETERMINED BY THE  $V_{CC}$  SUPPLY. THE OUTPUT FALL TIME IS LIMITED BY RC TIME CONSTANT OF THE LOAD.

Typical Performance Curves (Continued)

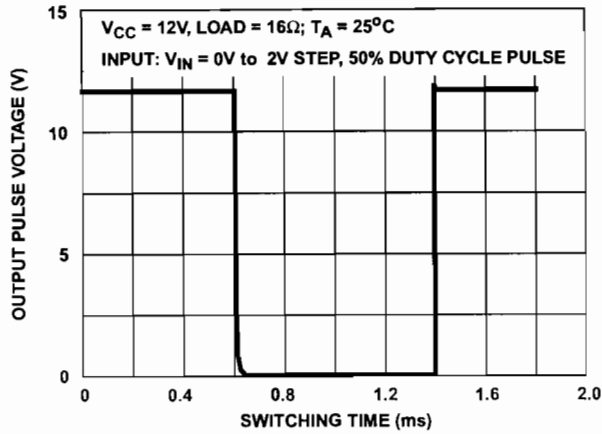


FIGURE 4. TYPICAL SWITCHING CHARACTERISTIC OF THE HIP1030 WITH AN OUTPUT RESISTIVE LOAD

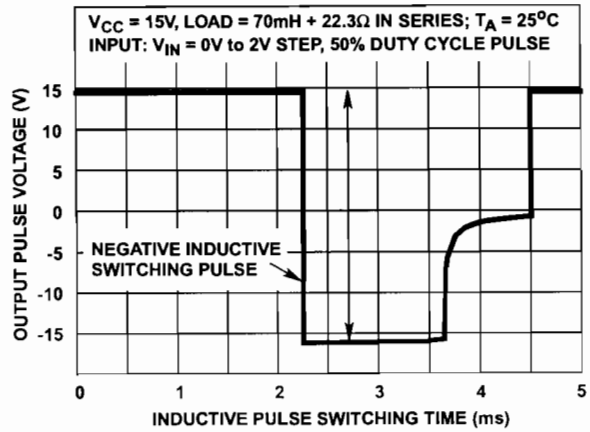


FIGURE 5. TYPICAL OUTPUT INDUCTIVE LOAD SWITCHING PULSE. THE NEGATIVE CLAMP VOLTAGE ( $V_{CC} - 31V$ ) FOR THE INDUCTIVE KICK PULSE IS REFERENCED TO THE  $V_{CC}$  SUPPLY INPUT

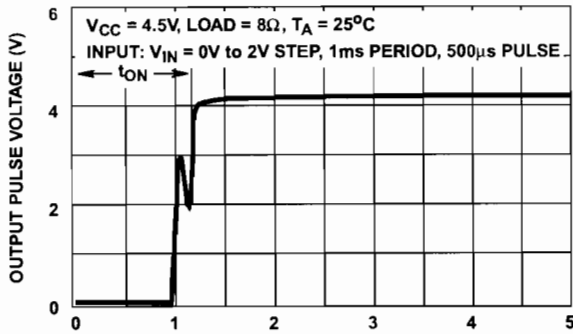


FIGURE 6A. TURN-ON TIME ( $\mu s$ )

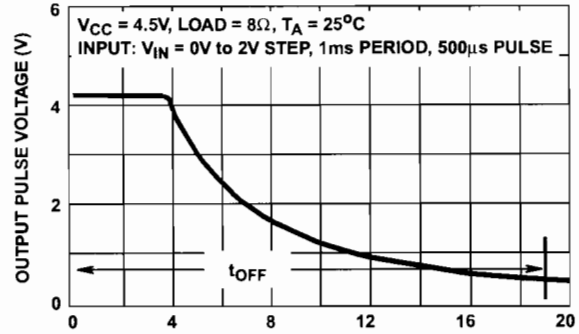


FIGURE 6B. TURN-OFF TIME ( $\mu s$ )

FIGURE 6. TYPICAL LOW SUPPLY VOLTAGE SWITCHING CHARACTERISTICS OF THE HIP1030. THE TURN-ON AND TURN-OFF CHARACTERISTICS ARE SHOWN FOR  $V_{CC} = 4.5V$ .

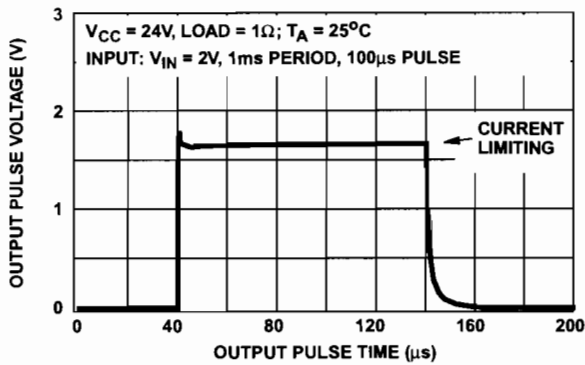


FIGURE 7. TYPICAL OUTPUT CURRENT PULSE WHEN SWITCHING INTO A LOW IMPEDANCE ( $1\Omega$ ), OR SHORTED LOAD. FOR THE CONDITIONS SHOWN, OUTPUT CURRENT LIMITING IS  $\sim 1.7A$

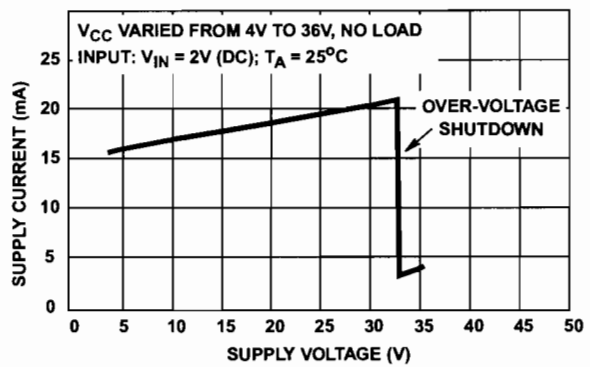
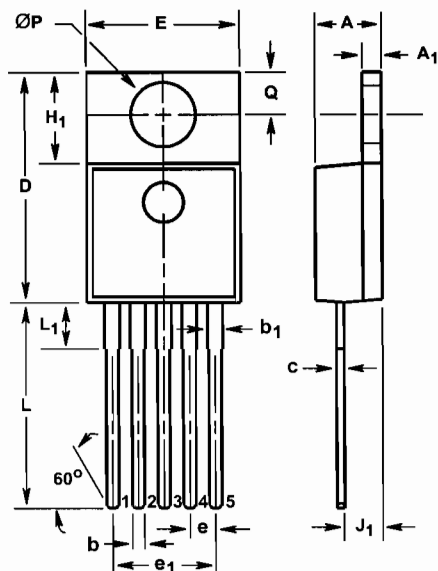


FIGURE 8. TYPICAL IDLE CURRENT vs SUPPLY VOLTAGE WITH NO LOAD

## Single-In-Line Plastic Packages (SIP)



**Z5.067C (ALTERNATE VERSION)**  
**5 LEAD PLASTIC SINGLE-IN-LINE PACKAGE**

SYMBOL	INCHES		MILLIMETERS		NOTES
	MIN	MAX	MIN	MAX	
A	0.170	0.180	4.32	4.57	-
A <sub>1</sub>	0.048	0.052	1.22	1.32	3, 4
b	0.030	0.034	0.77	0.86	3, 4
b <sub>1</sub>	0.031	0.041	0.79	1.04	3, 4
c	0.018	0.022	0.46	0.55	3, 4
D	0.590	0.610	14.99	15.49	-
E	0.395	0.405	10.04	10.28	-
e	0.067 TYP		1.70 TYP		5
e <sub>1</sub>	0.268 BSC		6.80 BSC		5
H <sub>1</sub>	0.235	0.255	5.97	6.47	-
J <sub>1</sub>	0.095	0.105	2.42	2.66	6
L	0.530	0.550	13.47	13.97	-
L <sub>1</sub>	0.110	0.130	2.80	3.30	2
ØP	0.149	0.153	3.79	3.88	-
Q	0.105	0.115	2.66	2.92	-

Rev. 1 4/96

## NOTES:

1. These dimensions are within allowable dimensions of Rev. A of JEDEC TS-001AA outline dated 8-89.
2. Solder finish uncontrolled in this area.
3. Lead dimension (without solder).
4. Add typically 0.002 inches (0.05mm) for solder plating.
5. Position of lead to be measured 0.250 inches (6.35mm) from bottom of dimension D.
6. Position of lead to be measured 0.100 inches (2.54mm) from bottom of dimension D.
7. Controlling dimension: Inch.

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